Claims

1. A space-time encoding and decoding method for a frequency selective fading channel, comprising:

A. an encoder taking two independent data fields of a time slot in input data as a processing unit with space-time orthogonal encoding method, encoding them and generating two data vectors, thereby forming two diversity signals, and transmitting said two diversity signals simultaneously with each through one diversity antenna;

B. a terminal receiving said two diversity signals, and neglecting mutual interference between said two diversity signals caused by non-orthogonality;

C. said terminal performing joint detection only taking into account affect to said two diversity signals from multipath interference and multi-user interference, thereby obtaining a decoding result; and

D. implementing interference counteraction based on result of joint diction to remove interference between two diversity signals, and then returning to step C to implement iteration for decoding processing.

2. The method of claim 1, wherein two diversity signals in step A are transmitted through two diversity beams of one smart antenna respectively and simultaneously.

3. The method of claim 1, further comprising the step of predefining number of iteration times to determine execution times from step C to step D and from step D to step C again.

4. The method of claim 1, wherein step B comprises: setting the upper right block and the lower left block of matrix $\mathbf{A}^{*T}\mathbf{A} = \begin{bmatrix} \mathbf{A}_1^{*T}\mathbf{A}_1 + \left(\mathbf{A}_2^{*T}\mathbf{A}_2\right)^* & \left(\mathbf{A}_1^{*T}\mathbf{A}_2\right)^T - \mathbf{A}_1^{*T}\mathbf{A}_2 \\ \left(\left(\mathbf{A}_1^{*T}\mathbf{A}_2\right)^T - \mathbf{A}_1^{*T}\mathbf{A}_2\right)^* & \left(\mathbf{A}_1^{*T}\mathbf{A}_1 + \left(\mathbf{A}_2^{*T}\mathbf{A}_2\right)^*\right)^* \end{bmatrix} \text{ to be null matrixes, and then }$

calculating equation $\hat{\mathbf{d}}_t = (\mathbf{B})^{-1} \mathbf{A}^{*T} \mathbf{r}$ to obtain a simplified equation for joint detection; wherein \mathbf{A}_1 and \mathbf{A}_2 are system matrixes of signal transmission between first and second transmitting antennas and receiving antennas; \mathbf{A} and \mathbf{B} are matrixes; $\hat{\mathbf{d}}_t$ is a value of continuous estimation of a receiving data field; \mathbf{r} is a sample value of said receiving data field; \mathbf{T} denotes a transpose operation; * denotes conjugate;

said joint detection in step C is calculated based on a simplified joint detection

equation:
$$\begin{cases} \hat{\mathbf{d}}(1) &= \mathbf{B}_{S}^{-1} \left(\mathbf{A}_{1}^{*T} \mathbf{r}_{1} + \left(\mathbf{A}_{2}^{*T} \mathbf{r}_{2} \right)^{*} \right) \\ \hat{\mathbf{d}}(2) &= \mathbf{B}_{S}^{-1} \left(\mathbf{A}_{1}^{*T} \mathbf{r}_{2} - \left(\mathbf{A}_{2}^{*T} \mathbf{r}_{1} \right)^{*} \right), \text{ wherein } \hat{\mathbf{d}}(1) \text{ and } \hat{\mathbf{d}}(2) \text{ are values} \end{cases}$$

of continuous estimation of two receiving data fields, \mathbf{B}_s is a matrix; \mathbf{r}_1 and \mathbf{r}_2 are sample values of two receiving data fields;

the step of implementing interference counteraction based on result of joint diction in step D further comprising:

D1. subtracting affect of a data field d(1) from received data signal based on the following formula,

$$\begin{cases} \mathbf{r}_1' &= \mathbf{r}_1 - \mathbf{A}_1 \widehat{\mathbf{d}}(1) \\ \mathbf{r}_2' &= \mathbf{r}_2 - \mathbf{A}_2 \widehat{\mathbf{d}}^*(1) \end{cases}$$

thereby obtaining $\mathbf{r_1'}$ and $\mathbf{r_2'}$; subtracting affect of another data field $\mathbf{d(2)}$ from received data signal based on the following formula:

$$\begin{cases} \mathbf{r}_{1}'' = \mathbf{r}_{1} + \mathbf{A}_{2} \hat{\mathbf{d}}^{*}(2) \\ \mathbf{r}_{2}'' = \mathbf{r}_{2} - \mathbf{A}_{1} \hat{\mathbf{d}}(2) \end{cases}$$

thereby obtaining $\mathbf{r_1''}$ and $\mathbf{r_2''}$;

- D2. substituting \mathbf{r}_1' and \mathbf{r}_2' for \mathbf{r}_1 and \mathbf{r}_2 in the second equation of said simplified joint detection formula used in step C, and substituting \mathbf{r}_1'' and \mathbf{r}_2'' for \mathbf{r}_1 and \mathbf{r}_2 in the first equation of said simplified joint detection formula used in step C, calculating said simplified joint detection formula, thereby obtaining iteration results of $\hat{\mathbf{d}}(1)$ and $\hat{\mathbf{d}}(2)$.
- 5. The method of claim 4, wherein said matrix **B** is calculated by one of the following formulas:

$$\mathbf{B} = \begin{cases} \mathbf{I} \\ \mathbf{A}^{*T} \mathbf{A} \\ \mathbf{A}^{*T} \mathbf{A} + \boldsymbol{\sigma}^{2} \mathbf{I} \end{cases}$$

these formulas including match filter scheme, zero-forcing block equalization scheme and minimum mean-square-error block equalization scheme; wherein σ^2 is noise power, and I is an identity matrix;

said matrix \mathbf{B}_{s} is calculated by one of the following formulas:

$$\mathbf{B}_{S} = \begin{cases} \mathbf{I} \\ \mathbf{A}_{1}^{*T} \mathbf{A}_{1} + \left(\mathbf{A}_{2}^{*T} \mathbf{A}_{2}\right)^{*} \\ \mathbf{A}_{1}^{*T} \mathbf{A}_{1} + \left(\mathbf{A}_{2}^{*T} \mathbf{A}_{2}\right)^{*} + \sigma^{2} \mathbf{I} \end{cases}$$

these formulas including match filter scheme, zero-forcing block equalization scheme and minimum mean-square-error block equalization scheme; wherein σ^2 is noise power, and I is an identity matrix.

6. The method of claim 4, wherein said system matrixes A_1 and A_2 are determined by channel pulse response and user transmission waveform.